

#### EEPROM PROGRAMMABLE VCXO CLOCK GENERATOR

#### **IDT5V19EE604**

### **Description**

The IDT5V19EE604 is a programmable clock generator intended for high performance data-communications, telecommunications, consumer, and networking applications. There are four internal PLLs, each individually programmable, allowing for four unique non-integer-related frequencies. The frequencies are generated from a single reference clock. The reference clock can come from one of the two redundant clock inputs. Automatic or manual switchover function allows any one of the redundant clocks to be selected during normal operation.

The IDT5V19EE604 is in-system, programmable and can be programmed through the use of I<sup>2</sup>C interface. An internal EEPROM allows the user to save and restore the configuration of the device without having to reprogram it on power-up.

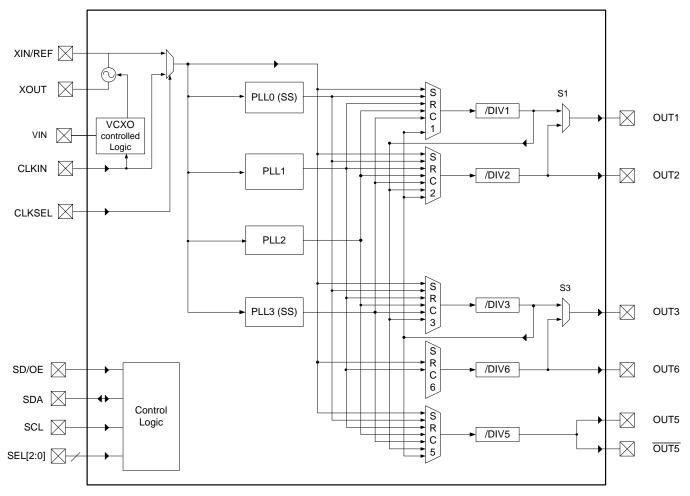
Each of the four PLLs has an 7-bit reference divider and a 12-bit feedback divider. This allows the user to generate four unique non-integer-related frequencies. The PLL loop bandwidth is programmable to allow the user to tailor the PLL response to the application. For instance, the user can tune the PLL parameters to minimize jitter generation or to maximize jitter attenuation. Spread spectrum generation and/or fractional divides are allowed on two of the PLLs.

There are a total of five 8-bit output dividers. The outputs are connected to the PLLs via a switch matrix. The switch matrix allows the user to route the PLL outputs to any output bank. This feature can be used to simplify and optimize the board layout. In addition, each output's slew rate and enable/disable function is programmable.

#### **Features**

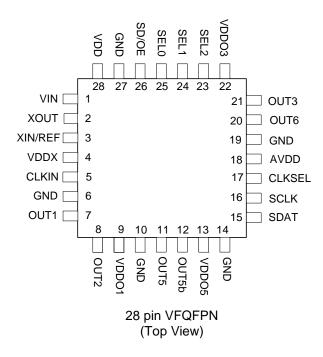
- Four internal PLLs
- Internal non-volatile EEPROM
- Fast (400kHz) mode I<sup>2</sup>C serial interface
- Input frequency range: 1 MHz to 200 MHz
- Output frequency range: 4.9 kHz to 200 MHz
- Reference crystal input with programmable linear load capacitance
  - Crystal frequency range: 8 MHz to 50 MHz
- Integrated VCXO
- Three independently controlled VDDO (1.8V 3.3V)
- Each PLL has a 7-bit reference divider and a 12-bit feedback-divider
- 8-bit output-divider blocks
- Fractional division capability on one PLL
- Two of the PLLs support spread spectrum generation capability
- I/O Standards:
  - Outputs 1.8 3.3 V LVTTL/ LVCMOS
  - Inputs 3.3 V LVTTL/ LVCMOS
- Programmable slew rate control
- Programmable loop bandwidth
- Programmable output inversion to reduce bimodal jitter
- Redundant clock inputs with auto and manual switchover options
- Individual output enable/disable
- Power-down mode
- 3.3V core V<sub>DD</sub>
- Available in VFQFPN package
- -40 to +85°C Industrial Temp operation

# **Functional Block Diagram**



1. CLKIN, CLKSEL, SD/OE and SEL[2:0] have pull down resistors.

# **Pin Configuration**



# **Pin Descriptions**

Pin#	Pin Name	I/O	Pin Type	Pin Description
1	VIN	I	LVTTL	VCXO analog control voltage input. Pulls output ±100ppm by varying from 0V to 3.3V.
2	XOUT	0	LVTTL	CRYSTAL_OUT Reference crystal feedback.
3	XIN / REF	I	LVTTL	CRYSTAL_IN Reference crystal input or external reference clock input.
4	VDDx		Power	Crystal oscillator power supply. Connect to 3.3V through $5\Omega$ resistor. Use filtered analog power supply if available.
5	CLKIN	I	LVTTL	Input clock. Weak internal pull down resistor.
6	GND		Power	Connect to Ground.
7	OUT1	0	LVTTL	Configurable clock output 1. Output levels controlled by VDDO1.
8	OUT2	0	LVTTL	Configurable clock output 2. Output levels controlled by VDDO1.
9	VDDO1		Power	Device power supply. Connect to 1.8 to 3.3V. Sets output voltage levels for OUT1 and OUT2.
10	GND		Power	Connect to Ground.
11	OUT5	0	LVTTL <sup>1</sup>	Configurable clock output 5. Output levels controlled by VDDO5.

Pin#	Pin Name	I/O	Pin Type	Pin Description
12	OUT5b	0	LVTTL <sup>1</sup>	Configurable clock output 5b. Output levels controlled by VDDO5.
13	VDDO5		Power	Device power supply. Connect to 1.8 to 3.3V. Sets output voltage levels for OUT5 and OUT5b.
14	GND		Power	Connect to Ground.
15	SDAT	I/O	Open Drain	Bidirectional I <sup>2</sup> C data. An external pull-up resistor is required. See I <sup>2</sup> C specification for pull-up value recommendation.
16	SCLK	I	LVTTL	I <sup>2</sup> C clock. An external pull-up resistor is required. See I <sup>2</sup> C specification for pull-up value recommendation.
17	CLKSEL	I	LVTTL	Input clock selector. Weak internal pull down resistor.
18	AVDD		Power	Device analog power supply. Connect to 3.3V. Use filtered analog power supply if available.
19	GND		Power	Connect to Ground.
20	OUT6	0	LVTTL	Configurable clock output 6. Output levels controlled by VDDO3.
21	OUT3	0	LVTTL	Configurable clock output 3. Output levels controlled by VDDO3.
22	VDDO3		Power	Device power supply. Connect to 1.8 to 3.3V. Sets output voltage levels for OUT3 and OUT6.
23	SEL2	I	LVTTL	Configuration select pin. Weak internal pull down resistor.
24	SEL1	I	LVTTL	Configuration select pin. Weak internal pull down resistor.
25	SEL0	I	LVTTL	Configuration select pin. Weak internal pull down resistor.
26	SD/OE	I	LVTTL	Enables/disables the outputs or powers down the chip. The SP bit (0x02) controls the polarity of the signal to be either active HIGH or LOW. (Default is active LOW.) Weak internal pull down resistor.
27	GND		Power	Connect to Ground.
28	VDD		Power	Device power supply. Connect to 3.3V.

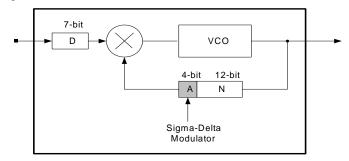
<sup>1.</sup> When only an individual single-ended clock output is required, tie OUT# and OUT#b together.

<sup>2.</sup> Analog power plane should be isolated from a 3.3V power plane through a ferrite bead.

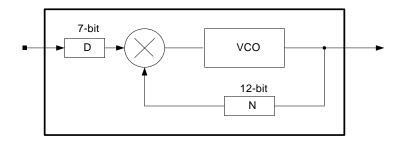
3. Each power pin should have a dedicated 0.01µF de-coupling capacitor. Digital VDDs may be tied together.

4. Unused clock inputs (REFIN or CLKIN) must be pulled high or low - they cannot be left floating. If the crystal oscillator is not used, XOUT must be left floating.

# **PLL Features and Descriptions**



**PLL0 Block Diagram** 



PLL1, PLL2 and PLL3 Block Diagram

	Pre-Divider (D) <sup>1</sup> Values	Multiplier (M) <sup>2</sup> Values	Programmable Loop Bandwidth	Spread Spectrum Generation Capability
PLL0	1 - 127	10 - 8206	Yes	Yes
PLL1	1 - 127	1 - 4095	Yes	No
PLL2	1 - 127	1 - 4095	Yes	No
PLL3	3 - 127	12 - 4095	Yes	Yes

<sup>1.</sup>For PLL0, PLL1 and PLL2, D=0 means PLL power down. For PLL3, 0, 1, and 2 are DNU (do not use) 2.For PLL0,  $M=2^*N+A+1$  (for A>0);  $M=2^*N$  (for A=0);  $A\le N-1$ . For PLL1, PLL2 and PLL3, A=N-10.

# Reference Clock Input Pins and Selection

The IDT5V19EE604 supports up to two clock inputs. One of the clock inputs (XIN/ REF) can be driven by either an external crystal or a reference clock. The second clock input (CLKIN) can only be driven from an external reference clock. The CLKSEL pin selects the input clock from either XTAL/REF or CLKIN.

Either clock input can be set as the primary clock. The primary clock designation is to establish which is the main reference clock to the PLLs. The non-primary clock is designated as the secondary clock in case the primary clock goes absent and a backup is needed. The PRIMSRC bit (0xBE through 0xC3) determines which clock input will be selected as primary clock. When PRIMSRC bit is "0", XIN/REF is selected as the primary clock, and when "1", CLKIN as the primary clock.

The two external reference clocks can be manually selected using the CLKSEL pin. The SM bits (0xBE through 0xC3) must be set to "0x" for manual switchover which is detailed in SWITCHOVER MODES section.

### Crystal Input (XIN/REF)

The crystal used should be a fundamental mode quartz crystal; overtone crystals should not be used.

When the XIN/REF pin is driven by a crystal, it is important to set the internal inverter oscillator drive strength and tuning/load capacitor values correctly to achieve the best clock performance. These values are programmable through I<sup>2</sup>C interface to allow for maximum compatibility with crystals from various manufacturers, processes, performances, and qualities. The internal load capacitors are true parallel-plate capacitors for ultra-linear performance. Parallel-plate capacitors were chosen to reduce the frequency shift that occurs when non-linear load capacitance interacts with load, bias, supply, and temperature changes. External non-linear crystal load capacitors should not be used for applications that are sensitive to absolute frequency requirements. The value of the internal load capacitors are determined by XTAL[4:0] bits. The load capacitance can be set with a resolution of 0.125 pF for a total crystal load ranging from 3.5 pF to 7.5 pF. Check with the crystal vendor's load capacitance specification for the exact setting to tune the internal load capacitor. The following equation governs how the total

internal load capacitance is set.

XTAL load cap = 3.5 pF + XTAL[4:0] \* 0.125 pF (Eq. 1)

Parameter	Bits	Step (pF)	Min (pF)	Max (pF)	
XTAL	8	0.125	0	4	

When using an external reference clock instead of a crystal on the XTAL/REF pin, the input load capacitors may be completely bypassed. This allows for the input frequency to be up to 200 MHz. When using an external reference clock, the XOUT pin must be left floating, XTAL must be programmed to the default value of "00h", and the crystal drive strength bit, XDRV (0x06), must be set to the default value of "11h".

#### **Switchover Modes**

The IDT5V19EE604 features redundant clock inputs which supports both Automatic and Manual switchover mode. These two modes are determined by the configuration bits, SM (0xBE through 0xC3). The primary clock source can be programmed, via the PRIMSRC bit, to be either XIN/REF or CLKIN. The other clock input will be considered as the secondary source. Note that the switchover modes are asynchronous. If the reference clocks are directly routed to OUTx with no phase relationship, short pulses can be generated during switchover. The automatic switchover mode will work only when the primary clock source is XIN/REF. Switchover modes are not supported for crystal input configurations.

#### **Manual Switchover Mode**

When SM[1:0] is "0x", the redundant inputs are in manual switchover mode. In this mode, CLKSEL pin is used to switch between the primary and secondary clock sources. As previously mentioned, the primary and secondary clock source setting is determined by the PRIMSRC bit. During the switchover, no glitches will occur at the output of the device, although there may be frequency and phase drift, depending on the exact phase and frequency relationship between the primary and secondary clocks.

#### **Automatic Switchover Mode**

The redundant inputs are in automatic switchover mode. Automatic switchover mode has revertive functionality. The input clock selection will switch to the secondary clock source when there are no transitions on the primary clock source for two secondary clock cycles. If both reference

clocks are at different frequencies, the device will always remain on the primary clock unless it is absent for two secondary clock cycles. The secondary clock must always run at a frequency less than or equal to the primary clock frequency.

# Reference Divider, Feedback Divider, and Output Divider

Each PLL incorporates a 7-bit reference divider (D[6:0]) and a 12-bit feedback divider (N[11:0]) that allows the user to generate four unique non-integer-related frequencies. Each output divide supports 8-bit output-divider (PM and Q[7:0]). The following equation governs how the output frequency is calculated.

$$F_{OUT} = \frac{F_{IN} * \left(\frac{M}{D}\right)}{ODIV}$$
 (Eq. 1)

Where FIN is the reference frequency, M is the total feedback-divider value, D is the reference divider value, ODIV is the total output-divider value, and FOUT is the resulting output frequency.

For PLLO,

$$M = 2 * N + A + 1 (for A>0)$$

$$M = 2 * N (for A = 0)$$

For PLL1, PLL2 and PLL3,

M = N

PM and Q[6:0] are the bits used to program the 8-bit output-dividers for outputs OUT1-6. OUT0 does not have any output divide along its path. The 8-bit output-dividers will bypass or divide down the output banks' frequency with even integer values ranging from 2 to 256.

There is the option to choose between disabling the output-divider, utilizing a div/1, a div/2, or the 7-bit Q-divider by using the PM bit. If the output is disabled, it will be driven High, Low or High Impedance, depending on OEM[1:0]. Each bank, except for OUT0, has a PM bit. When disabled, no clocks will appear at the output of the divider, but will remain powered on. The output divides selection table is shown below.

Q[6:0]	PM	Output Divider
111 1111	0	Disabled
	1	/1
<111 1111	0	/2
	1	/((Q[6:0] + 2) * 2)

Note that the actual 7-bit Q-divider value has a 2 added to the integer value Q and the outputs are routed through another div/2 block. The output divider should never be disabled unless the output bank will never be used during normal operation. The output frequency range are from 4.9KHz to 200MHz.

#### **Spread Spectrum Generation (PLL0)**

PLL0 supports spread spectrum generation capability, which users have the option of turning on or off. Spread spectrum profile, frequency, and spread amplitude are fully programmable. The programmable spread spectrum generation parameters are TSSC[3:0], NSSC[2:0], SS\_OFFSET[5:0], SD[3:0], DITH, and X2 bits. These bits are in the memory address from 0xAC to 0xBD for PLL0. The spread spectrum generation on PLL0 can be enabled/disabled using the TSSC[3:0] bits. To enable spread spectrum, set TSSC > '0' and set NSSC[2:0], SS\_OFFSET[5:0], SD[3:0], and the A[3:0] (in the total M value) accordingly. To disable spread spectrum generation, set TSSC = '0'.

#### TSSC[3:0]

These bits are used to determine the number of phase/frequency detector cycles per spread spectrum cycle (ssc) steps. The modulation frequency can be calculated with the TSSC bits in conjunction with the NSSC bits. Valid TSSC integer values for the modulation frequency range from 5 to 14. Values of 0 - 4 and 15 should not be used.

#### NSSC[2:0]

These bits are used to determine the number of delta-encoded samples used for a single quadrant of the spread spectrum waveform. All four quadrants of the spread spectrum waveform are mirror images of each other. The modulation frequency is also calculated based on the NSSC bits in conjunction with the TSSC bits. Valid NSSC integer values range from 1 to 6. Values of 0 and 7 should not be used.

#### SS\_OFFSET[5:0]

These bits are used to program the fractional offset with respect to the nominal M integer value. For center spread, the SS\_OFFSET is set to '0' so that the spread spectrum waveform is centered about the nominal M (Mnom) value. For down spread, the SS\_OFFSET > '0' such the spread spectrum waveform is centered about the (Mideal -1 +SS\_Offset) value. The downspread percentage can be thought of in terms of center spread. For example, a downspread of -1% can also be considered as a center spread of  $\pm 0.5\%$  but with Mnom shifted down by one and offset. The SS\_OFFSET has integer values ranging from 0 to 63.

#### SD[3:0]

These bits are used to shape the profile of the spread spectrum waveform. These are delta-encoded samples of the waveform. There are two sets of SD samples. The NSSC bits determine how many of these samples are repeated for the waveform. The sum of these delta-encoded samples (sigma delta-encoded samples) determine the amount of spread and should not exceed (63 - SS\_OFFSET). The maximum spread is inversely proportional to the nominal M integer value.

#### DITH

This bit is used for dithering the sigma-delta-encoded samples. This will randomize the least-significant bit of the input to the spread spectrum modulator. Set the bit to '1' to enable dithering.

#### **X2**

This bit will double the total value of the sigma-delta-encoded-samples which will increase the amplitude of the spread spectrum waveform by a factor of two. When X2 is '0', the amplitude remains nominal but if set to '1', the amplitude is increased by x2. The following equations govern how the spread spectrum is set:

$$Tssc = TSSC[3:0] + 2 (Eq. 2)$$

$$Nssc = NSSC[2:0] * 2 (Eq. 3)$$

$$SD[3:0]\kappa = S_{J+1}(unencoded) - S_{J}(unencoded)$$
 (Eq. 4)

where S<sub>J</sub> is the unencoded sample out of a possible 12 and SD<sub>K</sub> is the delta-encoded sample out of a possible 12.

Amplitude = 
$$((2*N[11:0] + A[3:0] + 1) * Spread% / 100) / 2$$
  
(Eq. 5)

if 1 < Amplitude < 2, then set X2 bit to '1'.

#### Modulation frequency:

$$FPFD = FIN / D (Eq. 6)$$

$$Fvco = Fpfd * Mnom (Eq. 7)$$

#### Spread:

$$\Sigma\Delta = SD_0 + SD_1 + SD_2 + ... + SD_{11}$$

the number of samples used depends on the Nssc value

$$\Sigma\Delta \leq 63$$
 - SS\_OFFSET

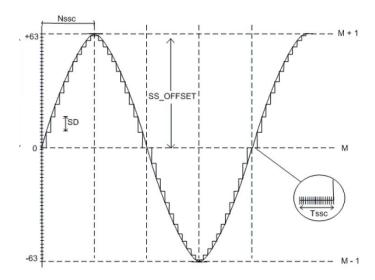
$$\pm$$
Spread% =  $(\Sigma \Delta * 100)/(64 * (2*N[11:0] + A[3:0] + 1) (Eq. 9)$ 

$$\pm$$
Max Spread% / 100 = 1 / Mnom or 2 / Mnom (X2=1)

### **Profile:**

Waveform starts with SS\_OFFSET, SS\_OFFSET + SD<sub>J</sub>, SS\_OFFSET + SD<sub>J+1</sub>, etc.

#### Spread Spectrum Using Sinusoidal Profile



#### **Example**

FIN = 25MHz, Fout = 100MHz, Fssc = 33KHz with center spread of  $\pm 2\%$ . Find the necessary spread spectrum register settings.

Since the spread is center, the SS\_OFFSET can be set to '0'. Solve for the nominal M value; keep in mind that the nominal M should be chosen to maximize

the VCO. Start with D = 1, using Eq.6 and Eq.7.

 $M_{NOM} = 1200MHz / 25MHz = 48$ 

Using Eq.4, we arbitrarily choose N = 22, A = 3. Now that we have the nominal M value, we can determine TSSC and NSSC by using Eq.8.

Nssc \* Tssc = 25MHz / (33KHz \* 4) = 190

However, using Eq. 2 and Eq.3, we find that the closest value is when TSSC = 14 and NSSC = 6. Keep in mind to maximize the number of samples used

to enhance the profile of the spread spectrum waveform.

$$Tssc = 14 + 2 = 16$$

$$Nssc = 6 * 2 = 12$$

Use Eq.10 to determine the value of the sigma-delta-encoded samples.

$$\pm 2\% = (\Sigma \Delta * 100)/(64 * 48)$$

$$\Sigma \Delta = 61.4$$

Either round up or down to the nearest integer value. Therefore, we end up with 61 or 62 for sigma-delta-encoded samples. Since the sigma-delta-encoded samples must not exceed 63 with SS\_OFFSET set to '0', 61 or 62 is well within the limits. It is the discretion of the user to define the shape of the profile that is better suited for the intended application.

Using Eq. 9 again, the actual spread for the sigma-delta-encoded samples of 56 and 57 are  $\pm 1.99\%$  and  $\pm 2.02\%$ , respectively.

Use Eq.10 to determine if the X2 bit needs to be set;

Therefore, the X2 = '0'. The dither bit is left to the discretion of the user.

The example above was of a center spread using spread spectrum. For down spread, the nominal M value can be set one integer value lower to 47.

Note that the IDT5V19EE604 should not be programmed with TSSC > '0', SS\_OFFSET = '0', and SD = '0' in order to prevent an unstable state in the modulator.

The PLL loop bandwidth must be at least 10x the modulation frequency along with higher damping (larger  $\omega z$ ) to prevent the spread spectrum from being filtered and reduce extraneous noise. Refer to the LOOP FILTER section for more detail on  $\omega z$ . The A[3:0] must be used for spread spectrum, even if the total multiplier value is an even integer.

### Spread Spectrum Generation (PLL3)

PLL3 support spread spectrum generation capability, which users have the option of turning on and off. Spread spectrum profile, frequency, and spread are fully programmable (within limits). The technique is different from that used in PLL0. The programmable spread spectrum generation parameters are SS\_D3[7:0], SSVCO[15:0], SSENB, IP3[4:0] and RZ3[3:0] bits. These bits are in the memory address range of 0x4C to 0x85 for PLL3. The spread spectrum generation on PLL3 can be enabled/disabled using the SSENB bit. To enable spread spectrum, set SSENB = '1'.

#### For Spread Enabled:

Spread spectrum is configured using SS\_D3(spread spectrum reference divide)

$$SS_D3 = \frac{F_{IN}}{4 * F_{MOD}}$$
 (Eq. 10)

and SSVCO (spread spectrum loop feedback counter).

SSVCO = 
$$[0.5 * \frac{F_{VCO}}{F_{MOD}} * (1 + SS/400) + 5]$$
 (Eq. 11)

SS is the total Spread Spectrum amount (I.e. center spread  $\pm 0.5\%$  has a total spread of 1.0% and down spread -0.5% has a total spread of 0.5%.)

#### **Loop Filter**

The loop filter for each PLL can be programmed to optimize the jitter performance. The low-pass frequency response of the PLL is the mechanism that dictates the jitter transfer characteristics. The loop bandwidth can be extracted from the jitter transfer. A narrow loop bandwidth is good for jitter attenuation while a wide loop bandwidth is best for low-jitter frequency generation. The specific loop filter components that can be programmed are the resistor via the RZ[3:0] bits, zero capacitor via the CZ bit (for PLL0, PLL1 and PLL2), and the charge pump current via the IP[2:0] bits (for PLL0, PLL1 and PLL2) or IP[3:0] (for PLL3).

The following equations govern how the loop filter is set for PLL0 - PLL2:

Resistor (Rz) = (RZ[0] + 2\*RZ[1]+4\*RZ[2] + 8\*RZ[3])\*4.0 kOhm

Pole capacitor (Cp) = 15 pF

Charge pump (Ip) = 
$$6 * (IP[0] + 2*IP[1]+4*IP[2]) uA$$

VCO gain (Kvco) = 900 MHz/V \* 
$$2\pi$$

The following equations govern how the loop filter is set for PLL3:

For Non-Spread Spectrum Operation:

Resistor(Rz) = 
$${12.5 + 12.5*(RZ[1] + 2*RZ[2] + 4*RZ[3])) \over * RZ[0] + 6*(1 - RZ[0])}$$
 kOhms (Eq. 12)

For Spread Spectrum Operation:

$$Resistor(Rz) = {(62.5 + 12.5*(RZ[1] + 2*RZ[2] + 4*RZ[3])) \atop * RZ[0] + 6*(1 - RZ[0])} kOhms (Eq. 13)$$

Zero capacitor (Cz) = 250 pF

Pole capacitor (Cp) = 15 pF

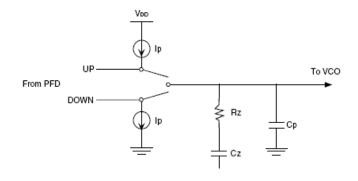
For Non-Spread Spectrum Operation:

$$\frac{\text{Charge pump (lp)}}{\text{pump (lp)}} = \frac{24*(1+(2*IP[0])+(4*IP[1])+(8*IP[2]))}{3+(5*IP[3])+(11*IP[4])} \quad \text{A (Eq. 14)}$$

For Spread Spectrum Operation:

$$\begin{array}{ll} \text{Charge} & = \frac{12 * (1 + (2 * IP[0]) + (4 * IP[1]) + (8 * IP[2]))}{27 + (5 * IP[3]) + (11 * IP[4])} \ \ \, \text{A} \ \ \, \text{(Eq. 14)} \\ \end{array}$$

VCO gain (Kyco) = 900 MHz/V \*  $2\pi$ 



#### **PLL Loop Bandwidth:**

Charge pump gain  $(K\phi) = Ip / 2\pi$ 

VCO gain (Kvco) = 900 MHz/V \*  $2\pi$ 

M = Total multiplier value (See the Reference Divider, Feedback Divider and Output Divider section for more detail)

$$\omega c = (Rz * K\phi * Kvco * Cz)/(M * (Cz + Cp))$$

$$Fc = \omega c / 2\pi$$

Note, the phase/frequency detector frequency (FPFD) is typically seven times the PLL closed-loop bandwidth (Fc) but too high of a ratio will reduce the phase margin thus compromising loop stability.

To determine if the loop is stable, the phase margin (φm) needs to be calculated as follows.

#### **Phase Margin:**

$$\omega z = 1 / (Rz * Cz)$$

$$\omega p = (Cz + Cp)/(Rz * Cz * Cp)$$

$$\phi m = (360 / 2\pi) * [tan_{-1}(\omega c / \omega z) - tan_{-1}(\omega c / \omega p)]$$

To ensure stability in the loop, the phase margin is recommended to be  $> 60^{\circ}$  but too high will result in the lock time being excessively long. Certain loop filter parameters would need to be compromised to not only meet a required loop bandwidth but to also maintain loop stability.

# SEL[2:0] Function

The IDT5V19EE604 can support up to six unique configurations. Users may pre-programmed all these configurations, and select the configurations using SEL[2:0] pins. Alternatively, users may use I<sup>2</sup>C interface to configure these registers on-the-fly.

SEL2	SEL1	SEL0	Configuration Selections
0	0	0	Select CONFIG0
0	0	1	Select CONFIG1
0	1	0	Select CONFIG2
0	1	1	Select CONFIG3
1	0	0	Select CONFIG4
1	0	1	Select CONFIG5
1	1	0	Reserved (Do not use)
1	1	1	Reserved (Do not use)

### **Crystal/Clock Selection**

XTCLKSEL bit is used to bypass a crystal oscillator circuit when external clock source is used.

PRIMSRC bit is used to select a primary clock from XIN/REF and CLKIN.

PRIMSRC bit	Primary	Secondary
0	XIN/REF	CLKIN
1	CLKIN	XIN/REF

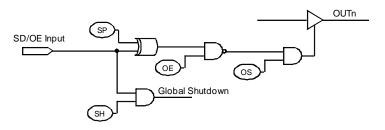
<b>CLKSEL</b> input	Clock Source
0	Primary Clock Source
1	Secondary Clock Source

CLKSEL	PRIMSRC	Reference Clock
0	0	XIN/REF
0	1	CLKIN
1	0	CLKIN
1	1	XIN/REF

SMx[1:0]	Swithcing Mode	Primary to Secondary	Secondary to Primary
0x	Manual	No	No
10	Auto	Yes	No
11	Auto-Revertive	Yes	Yes

#### **SD/OE Pin Function**

The polarity of the SD/OE signal pin can be programmed to be either active HIGH or LOW with the SP bit (0x02). When SP is "0" (default), the pin becomes active LOW and when SP is "1", the pin becomes active HIGH. The SD/OE pin can be configured as either to shutdown the PLLs or to enable/disable the outputs.



**Truth Table** 

SH bit	SP bit	OSn bit	OEn bit	SD/OE	OUTn
0	0	0	Х	Х	High-Z <sup>2</sup>
0	0	1	0	Х	Enabled
0	0	1	1	0	Enabled
0	0	1	1	1	Suspended
0	1	0	Х	Х	High-Z <sup>2</sup>
0	1	1	0	Х	Enabled
0	1	1	1	0	Suspended
0	1	1	1	1	Enabled
1	0	0	Х	0	High-Z <sup>2</sup>
1	0	1	0	0	Enabled
1	0	1	1	0	Enabled
1	1	0	Х	0	High-Z <sup>2</sup>
1	1	1	0	0	Enabled
1	1	1	1	0	Suspended
1	Х	Х	Х	1	Suspended <sup>1</sup>

Note 1: Global Shutdown

Note 2: Hi-Z regardless of OEM bits

### **Configuration OUTx IO Standard**

Users can configure the individual output IO standard from a specified 1.8 to 3.3V power supplies. Each output can support 1.8 to 3.3V LVTTL. OUT0 can only be a 3.3V single-ended output.

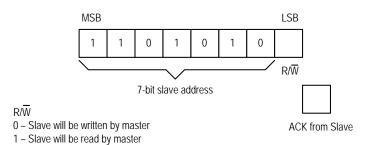
#### **Programming the Device**

I<sup>2</sup>C may be used to program the IDT5V19EE604.

- Device (slave) address = 7'b1101010

### I<sup>2</sup>C Programming

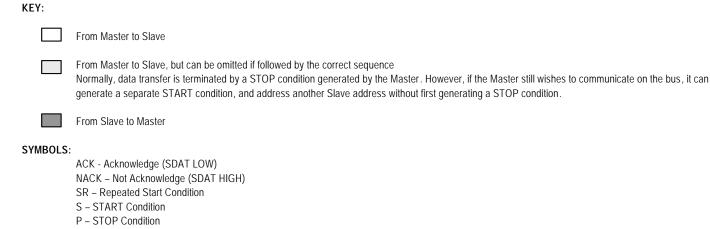
The IDT5V19EE604 is programmed through an I<sup>2</sup>C-Bus serial interface, and is an I<sup>2</sup>C slave device. The read and write transfer formats are supported. The first byte of data after a write frame to the correct slave address is interpreted as the register address; this address auto-increments after each byte written or read.



The first byte transmitted by the Master is the Slave Address followed by the  $R/\overline{W}$  bit. The Slave acknowledges by sending a "1" bit.

## First Byte Transmitted on I<sup>2</sup>C Bus

#### External I<sup>2</sup>C Interface Condition



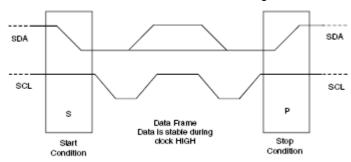
#### **Progwrite**

S	Address	R/W	ACK	Command Code	ACK	Register	ACK	Data	ACK	Р
	7-bits	0	1-bit	8-bits: xxxx xx00	1-bit	8-bits	1-bit	8-bits	1-bit	

#### **Progwrite Command Frame**

Writes can continue as long as a Stop condition is not sent and each byte will increment the register address.

#### The frame formats are shown in the following illustration.



Framing

#### **Progread**

Note: If the expected read command is not from the next higher register to the previous read or write command, then set a known "read" register address prior to a read operation by issuing the following command:

S	Address	R/W	ACK	Command Code	ACK	Register	ACK	Р
	7-bits	0	1-bit	8-bits: xxxx xx00	1-bit	8-bits	1-bit	

Prior to Progread Command Set Register Address

The user can ignore the STOP condition above and use a repeated START condition instead, straight after the slave acknowledgement bit (i.e., followed by the Progread command):

S	Address	R/W	ACK	ID Byte	ACK	Data_1	ACK	Data_2	ACK	Data_last	NACK	Р
	7-bits	1	1-bit	8-bits	1-bit	8-bits	1-bit	8-bits	1-bit	8-bits	1-bit	

#### **Progread Command Frame**

#### **Progsave**

S	Address	R/W	ACK	<b>Command Code</b>	ACK	Р
	7-bits	0	1-bit	8-bits: xxxx xx01	1-bit	

Note:

PROGWRITE is for writing to the IDT5V19EE604 registers.

PROGREAD is for reading the IDT5V19EE604 registers.

PROGSAVE is for saving all the contents of the IDT5V19EE604 registers to the EEPROM.

PROGRESTORE is for loading the entire EEPROM contents to the IDT5V19EE604 registers.

#### **Progrestore**

S	Address	R/W	ACK	Command Code	ACK	Р
	7-bits	0	1-bit	8-bits: xxxx xx10	1-bit	

#### **EEPROM Interface**

The IDT5V19EE604 can also store its configuration in an internal EEPROM. The contents of the device's internal programming registers can be saved to the EEPROM by issuing a save instruction (ProgSave) and can be loaded back to the internal programming registers by issuing a restore instruction (ProgRestore).

To initiate a save or restore using I<sup>2</sup>C, only two bytes are transferred. The Device Address is issued with the read/write bit set to "0", followed by the appropriate command code. The save or restore instruction executes after the STOP condition is issued by the Master, during which time the IDT5V19EE604 will not generate Acknowledge bits. The IDT5V19EE604 will acknowledge the instructions after it has completed execution of them. During that time, the I<sup>2</sup>C bus should be interpreted as busy by all other users of the bus.

On power-up of the IDT5V19EE604, an automatic restore is performed to load the EEPROM contents into the internal programming registers. The IDT5V19EE604 will be ready to accept a programming instruction once it acknowledges its 7-bit I<sup>2</sup>C address.

# I<sup>2</sup>C Bus DC Characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IH</sub>	Input HIGH Level		0.7xV <sub>DD</sub>			V
$V_{IL}$	Input LOW Level				0.3xV <sub>DD</sub>	V
V <sub>HYS</sub>	Hysteresis of Inputs		0.05xV <sub>DD</sub>			V
I <sub>IN</sub>	Input Leakage Current				±1.0	μΑ
V <sub>OL</sub>	Output LOW Voltage	I <sub>OL</sub> = 3 mA			0.4	V

# I<sup>2</sup>C Bus AC Characteristics for Standard Mode

Symbol	Parameter	Min	Тур	Max	Unit
F <sub>SCLK</sub>	Serial Clock Frequency (SCL)	0		100	kHz
t <sub>BUF</sub>	Bus free time between STOP and START	4.7			μs
t <sub>SU:START</sub>	Setup Time, START	4.7			μs
t <sub>HD:START</sub>	Hold Time, START	4			μs
t <sub>SU:DATA</sub>	Setup Time, data input (SDA)	250			ns
t <sub>HD:DATA</sub>	Hold Time, data input (SDA) <sup>1</sup>	0			μs
t <sub>OVD</sub>	Output data valid from clock			3.45	μs
C <sub>B</sub>	Capacitive Load for Each Bus Line			400	pF
t <sub>R</sub>	Rise Time, data and clock (SDAT, SCLK)			1000	ns
t <sub>F</sub>	Fall Time, data and clock (SDAT, SCLK)			300	ns
t <sub>HIGH</sub>	HIGH Time, clock (SCLK)	4			μs
t <sub>LOW</sub>	LOW Time, clock (SCLK)	4.7			μs
t <sub>SU:STOP</sub>	Setup Time, STOP	4			μs

Note 1: A device must internally provide a hold time of at least 300 ns for the SDAT signal (referred to the  $V_{IH}(MIN)$  of the SCLK signal) to bridge the undefined region of the falling edge of SCLK.

# I<sup>2</sup>C Bus AC Characteristics for Fast Mode

Symbol	Parameter	Min	Тур	Max	Unit
F <sub>SCLK</sub>	Serial Clock Frequency (SCL)	0		400	kHz
t <sub>BUF</sub>	Bus free time between STOP and START	1.3			μs
t <sub>SU:START</sub>	Setup Time, START	0.6			μs
t <sub>HD:START</sub>	Hold Time, START	0.6			μs
t <sub>SU:DATA</sub>	Setup Time, data input (SDA)	100			ns
t <sub>HD:DATA</sub>	Hold Time, data input (SDA) <sup>1</sup>	0			μs
t <sub>OVD</sub>	Output data valid from clock			0.9	μs
C <sub>B</sub>	Capacitive Load for Each Bus Line			400	pF
t <sub>R</sub>	Rise Time, data and clock (SDA, SCL)	20 + 0.1xC <sub>B</sub>		300	ns
t <sub>F</sub>	Fall Time, data and clock (SDA, SCL)	20 + 0.1xC <sub>B</sub>		300	ns
t <sub>HIGH</sub>	HIGH Time, clock (SCL)	0.6			μs
t <sub>LOW</sub>	LOW Time, clock (SCL)	1.3			μs
t <sub>SU:STOP</sub>	Setup Time, STOP	0.6			μs

Note 1: A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the  $V_{IH}(MIN)$  of the SCL signal) to bridge the undefined region of the falling edge of SCL.

### **Absolute Maximum Ratings**

Stresses above the ratings listed below can cause permanent damage to the IDT5V19EE604. These ratings, which are standard values for IDT commercially rated parts, are stress ratings only. Functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods can affect product reliability. Electrical parameters are guaranteed only over the recommended operating temperature range.

Symbol	Description	Min	Max	Unit
V <sub>DD</sub>	Internal Power Supply Voltage	-0.5	+4.6	V
V <sub>I</sub>	Input Voltage <sup>1</sup>	-0.5	+4.6	V
V <sub>O</sub>	Output Voltage (not to exceed 4.6 V) <sup>1</sup>	-0.5	V <sub>DD</sub> +0.5	V
T <sub>J</sub>	Junction Temperature		150	°C
T <sub>STG</sub>	Storage Temperature	-65	150	°C

<sup>1.</sup>Input negative and output voltage ratings may be exceeded if the input and output current ratings are observed.

### **Recommended Operation Conditions**

Symbol	Parameter	Min	Тур	Max	Unit
V <sub>DD</sub>	Power supply voltage for V <sub>DD</sub> pins supporting core and outputs	3.135	3.3	3.465	V
V <sub>DDX</sub>	Power supply voltage for crystal oscillator. Use filtered analog power supply if available.	3.135	3.3	3.465	V
AV <sub>DD</sub>	Analog power supply voltage. Use filtered analog power supply if available.	3.135	3.3	3.465	V
$V_{DDOX}$	3.3V VDDO Range	3.0	3.3	3.6	V
	2.5V VDDO Range for 2.5V LVTTL	2.25	2.5	2.75	V
	1.8V VDDO Range for 1.8V LVTTL	1.7	1.8	1.9	V
	Power supply voltage for V <sub>DD</sub> pins supporting LVDS/LVPECL/HCSL outputs	3.135	3.3	3.465	V
T <sub>A</sub>	Operating temperature, ambient	-40		+85	°C
C <sub>LOAD_OUT</sub>	Maximum load capacitance (3.3V LVTTL only)			15	pF
	Maximum load capacitance (1.8V/2.5V LVTTL only)			8	pF
F <sub>IN</sub>	External reference crystal	8		50	MHz
	External reference clock CLKIN	1		200	
t <sub>PU</sub>	Power up time for all V <sub>DD</sub> s to reach minimum specified voltage (power ramps must be monotonic)	0.05		5	ms

# $\label{eq:capacitance} \textbf{Capacitance} \,\, (\textbf{T}_{A} = +25 \,\, ^{\circ}\text{C}, \, \textbf{f}_{IN} = 1 \,\, \text{MHz}, \, \text{VIN} = 0 \text{V})$

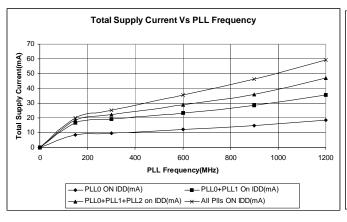
Symbol	Parameter	Min	Тур	Max	Unit
C <sub>IN</sub>	Input Capacitance (VIN, CLKIN, CLKSEL, SD/OE, SDA, SCL, SEL[2:0])		3	7	pF
Pull-down Resistor	CLKIN, CLKSEL, SD/OE, SEL[2:0]		180		kΩ
Crystal Specifi	cations				
XTAL_FREQ	Crystal frequency	8		50	MHz
XTAL_MIN	Minimum crystal load capacitance	3.5			pF
XTAL_MAX	Maximum crystal load capacitance			35.5	pF
XTAL_V <sub>PP</sub>	Voltage swing (peak-to-peak, nominal)	1.5	2.3	3.2	V

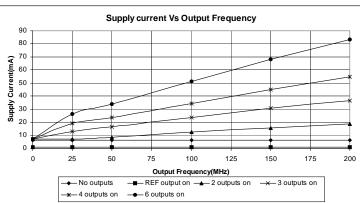
# DC Electrical Characteristics for 3.3-V LVTTL <sup>1</sup>

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
V <sub>OH</sub>	Output HIGH Voltage		2.4		$V_{DD}$	V
V <sub>OL</sub>	Output LOW Voltage				0.4	V
V <sub>IH</sub>	Input HIGH Voltage		2			V
V <sub>IL</sub>	Input LOW Voltage				0.8	V
I <sub>OZDD</sub>	Output Leakage Current	3-state outputs. $V_O = V_{DD}$ or GND, $V_{DD} = 3.6V$			10	μА
VIN	VCXO Control Voltage		0		3.3	V

Note 1: See "Recommended Operating Conditions" table.

# **Power Supply Characteristics for PLLs and LVTTL Outputs**





# **AC Timing Electrical Characteristics**

(Spread Spectrum Generation = OFF)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Units
f <sub>IN</sub> 1	Input Frequency	Input frequency limit (CLKIN)	1		200	MHz
		Input frequency limit (XIN/REF)	8		100	MHz
1 / t1	Output Frequency	Single ended clock output limit	0.001		200	MHz
f <sub>VCO</sub>	VCO Frequency	VCO operating frequency range	100		1200	MHz
f <sub>PFD</sub>	PFD Frequency	PFD operating frequency range	0.5 <sup>1</sup>		100	MHz
$f_{BW}$	Loop Bandwidth	Based on loop filter resistor and capacitor values	0.01		10	MHz
t2	Input Duty Cycle	Duty Cycle for input	40		60	%
t3	Output Duty Cycle	Measured at V <sub>DD</sub> /2, all outputs except Reference output	45		55	%
		Measured at V <sub>DD</sub> /2, Reference output	40		60	%
t4 <sup>2</sup>	Slew Rate, SLEW[1:0] = 00	Single-ended 3.3V LVCMOS output clock rise and fall time, 20% to 80% of V <sub>DD</sub> (Output Load = 5 pF)		3.5		V/ns
	Slew Rate, SLEW[1:0] = 01	Single-ended 3.3V LVCMOS output clock rise and fall time, 20% to 80% of V <sub>DD</sub> (Output Load = 5 pF)		2.75		
	Slew Rate, SLEW[1:0] = 10	Single-ended 3.3V LVCMOS output clock rise and fall time, 20% to 80% of V <sub>DD</sub> (Output Load = 5 pF)		2		
	Slew Rate, SLEW[1:0] = 11	Single-ended 3.3V LVCMOS output clock rise and fall time, 20% to 80% of V <sub>DD</sub> (Output Load = 5 pF)		1.25		
t7	Clock Jitter	Peak-to-peak period jitter, 1PLL, multiple output frequencies switching		80	100	ps
		Peak-to-peak period jitter, all 4 PLLs on <sup>3</sup>		200	270	ps
t8	Output Skew	Skew between output to output on the same bank			75	ps
t9 <sup>4</sup>	Lock Time	PLL lock time from power-up		10	20	ms
t10 <sup>5</sup>	Lock Time	PLL lock time from shutdown mode			2	ms
K <sub>VCXO</sub>	VCXO Gain	$VIN = V_{DD}/2 \pm 1V$		75	100	ppm/V
	Crystal Pullability <sup>6</sup>	0V ≤ VIN ≤ 3.3V	-100		100	ppm

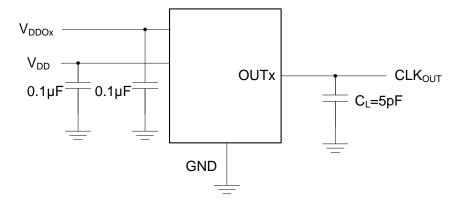
<sup>1.</sup> Practical lower frequency is determined by loop filter settings.
2. A slew rate of 2.75 V/ns or greater should be selected for output frequencies of 100MHz or higher.
3. Jitter measured with clock outputs of 27 MHz, 48 MHz, 24.576 MHz, 74.25 MHz and 25 MHz.
4. Includes loading the configuration bits from EEPROM to PLL registers. It does not include EEPROM programming/write time.
5. Actual PLL lock time depends on the loop configuration.
6. With a pullable crystal that conforms to IDT's specifications.

# **Spread Spectrum Generation Specifications**

Symbol	Parameter	Description	Min	Тур	Max	Unit
f <sub>IN</sub> 1	Input Frequency	Input Frequency Limit	1		400	MHz
$f_{MOD}$	Mod Frequency	Modulation Frequency		33	120	kHz
f <sub>SPREAD</sub>	Spread Value	Amount of Spread Value (programmable) - Down Spread	-0.5		-4.0	%f <sub>OUT</sub>
		Amount of Spread Value (programmable) - Center Spread	±0.25		±2.0	

<sup>1.</sup> Practical lower frequency is determined by loop filter settings.

### **Test Circuits and Conditions**



**Test Circuits for DC Outputs** 

# **Programming Registers Table**

	Default				В	Bit #				
Addr	Register Hex Value	7	6	5	4	3	2	1	0	Description
0x00	00			1	Reserved		1	1	HW/SW	Hardware/Software Mode control HW/SW - 0=HW, 1=SW
0x01	00			Reserved				SEL[2:0]		SEL[2:0] - selects configuration in SW mode
0x02	02	SP	OE6	OE5	OE4	OE3	OE2	OE1	OE0	OEx=Output Power Suspend function for OUTx ('1'=OUTx will be suspended on SD/OE pin. Disable mode is defined by OEMx bits), '0'=outputs enabled and no association with OE pin (default).
0x03	02	Reserved		OS*[6:0]					OS*[6:0] - output suspend, active low. Overwrites OE setting.	
0x04	0F	SH		Reserved			PL	LS*[3:0]		PLLS*[3:0] - PLL Suspend, active low SH - shutdown/OE configuration
0x05	04	Reserved			XTCLKSEL					XTCLKSEL - crystal/clock select. 0=Crytal, 1=ICLK
0x06	00	Rese	Reserved XDRV[1:0]				Re	eserved		Crystal drive finetune XDRV[1:0] - crystal drive strength for VCXO
0x07	00		Reserved				XTAL[4:0]			XTAL[4:0] - crystal cap
0x08	00		GAII	N[3:0]			OFF	SET[3:0]		VCXO bits
0x09	00			Reserved						Reserved
0x0A	10	CZ0_CFG4		IP0[2:0]_CFG4				3:0]_CFG4		PLL0 loop parameter
0x0B	10	CZ0_CFG5		IP0[2:0]_CFG			-	3:0]_CFG5		
0x0C	10	CZ0_CFG0		IP0[2:0]_CFG(				3:0]_CFG0		
0x0D 0x0E	10 10	CZ0_CFG1 CZ0_CFG2		IP0[2:0]_CFG				3:0]_CFG1 3:0]_CFG2		_
0x0F	10	CZ0_CFG3		IP0[2:0]_CFG3			_	3:0]_CFG3		_
0x10	00	Reserved		11 0[2.0]_01 00	,	D0[6:0]_CFG		<u></u>		PLL0 input divider and input sell
0x11	00	Reserved				D0[6:0]_CF0				D0[6:0] - 127 step Ref Div
0x12	00	Reserved				D0[6:0]_CFG	i2			D0 = 0 means power down.
0x13	00	Reserved				D0[6:0]_CF0	i3			
0x14	00	Reserved				D0[6:0]_CFG	ì4			
0x15	00	Reserved				D0[6:0]_CF0	ì5			
0x16	01					0]_CFG4				N - Feedback Divider
0x17	01	_				0]_CFG5				2 - 4095 (values of "0" and "1" are not allowed) Total feedback with A,
0x18	01					0]_CFG0				using provided calculation
0x19	01					0]_CFG1				_
0x1A	01					0]_CFG2				_
0x1B 0x1C	01		V0[5:0	_CFG0	NU[7:	0]_CFG3	NOT4 4	:8]_CFG0		
0x1C 0x1D	00			_CFG0  _CFG1			-	:8]_CFG0 :8]_CFG1		_
0x1E	00			_CFG1  _CFG2				:8]_CFG2		_
0x1F	00			CFG3				:8]_CFG3		<del>- </del>
0x20	00		-	CFG4				:8]_CFG4		-
0x21	00			 L_CFG5			_	:8]_CFG5		-

	Default				В	Bit #						
Addr	Register Hex Value	7	6	5	4	3	2	1	0	Description		
0x22	10	CZ1_CFG4		IP1[2:0]_CFG4	l .		RZ1[3	B:0]_CFG4	<u> </u>	PLL1 Loop Parameter		
0x23	10	CZ1_CFG5	IP1[2:0]_CFG5 RZ1[3:0]_CFG5									
0x24	10	CZ1_CFG0		IP1[2:0]_CFG0			RZ1[3	3:0]_CFG0				
0x25	10	CZ1_CFG1		IP1[2:0]_CFG1			RZ1[3	3:0]_CFG1				
0x26	10	CZ1_CFG2		IP1[2:0]_CFG2			RZ1[3	3:0]_CFG2				
0x27	10	CZ1_CFG3		IP1[2:0]_CFG3								
0x28	00	Reserved				D1[6:0]_CFG	0			PLL1 input divider and input sel		
0x29	00	Reserved				D1[6:0]_CFG	1			D1[6:0] - 127 step Ref Div D1 = 0 means power down.		
0x2A	00	Reserved				D1[6:0]_CFG	2			D1 = 0 means power down.		
0x2B	00	Reserved				D1[6:0]_CFG	3					
0x2C	00	Reserved				D1[6:0]_CFG	4					
0x2D	00	Reserved				D1[6:0]_CFG	5					
0x2E	01				N1[7:	0]_CFG4				N - Feedback Divider		
0x2F	01				N1[7:	0]_CFG5				2 - 4095 (value of "0" is not allowed) Total feedback with A, using		
0x30	01				N1[7:	0]_CFG0				provided calculation		
0x31	01				N1[7:	0]_CFG1						
0x32	01				N1[7:	0]_CFG2						
0x33	01				N1[7:	0]_CFG3						
0x34	00	N3[11:8]_CFG0 N1[11:8]_CFG0							PLL3 Feedback Divider			
0x35	00		8]_CFG1			N1[11	:8]_CFG1					
0x36	00		N3[11:8]_CFG2 N1[11:8]_CFG2									
0x37	00		N3[11:8]_CFG3 N1[11:8]_CFG3									
0x38	00		N3[11:8	8]_CFG4			N1[11	:8]_CFG4				
0x39	00		N3[11:8	8]_CFG5			N1[11	:8]_CFG5				
0x3A	00	CZ2_CFG4		IP2[2:0]_CFG4			RZ2[3	3:0]_CFG4		PLL2 Loop Parameter		
0x3B	00	CZ2_CFG5		IP2[2:0]_CFG5			RZ2[3	3:0]_CFG5				
0x3C	00	CZ2_CFG0		IP2[2:0]_CFG0			RZ2[3	3:0]_CFG0				
0x3D	00	CZ2_CFG1		IP2[2:0]_CFG1		RZ2[3:0]_CFG1						
0x3E	00	CZ2_CFG2		IP2[2:0]_CFG2			RZ2[3	3:0]_CFG2				
0x3F	00	CZ2_CFG3		IP2[2:0]_CFG3		RZ2[3:0]_CFG3						
0x40	00	Reserved			•	D2[6:0]_CFG	0			PLL2 Reference Divide and Input		
0x41	00					D2[6:0]_CFG	1			Select D2[6:0] - 127 step Ref Div		
0x42	00					D2[6:0]_CFG	2			D2 = 0 means power down.		
0x43	00					D2[6:0]_CFG	3					
0x44	00					D2[6:0]_CFG	4					
0x45	00					D2[6:0]_CFG	5					
0x46	01				_	0]_CFG4				N2[7:0] - PLL2 Feedback Divider		
0x47	01				-	0]_CFG5				2 - 4095 (value of "0" is not allowed).		
0x48	01				N2[7:	0]_CFG0				(See Addr 0x4C:0x51 for N2[15:8])		
0x49	01				_	0]_CFG1						
0x4A	01					0]_CFG2						
0x4B	01		N2[7:0]_CFG3									
0x4C	80	SSENB_CFG0	0	0	IP3[4]_CFG0		_	:8]_CFG0		N2[11:8] - PLL2 Feedback Divide PLL3 Spread Spectrum SSENB - Spread Spectrum Enable		
0x4D	80	SSENB_CFG1	0	0	IP3[4]_CFG1			:8]_CFG1				
0x4E	80	SSENB_CFG2 0 0   1P3[4]_CFG2   N2[11:8]_CFG2   SSENB = 1 means OI					SSENB = 1 means ON					
0x4F	80	SSENB_CFG3	Cumont						IP3[4:0] - PLL3 Charge Pump			
0x50	80	SSENB_CFG4	0	0	IP3[4]_CFG4	-		:8]_CFG4	-	Current.		
0x51	80	SSENB_CFG5	SENB_CFG5 0 0 IP3[4]_CFG5 N2[11:8]_CFG5									
0x52	XX <sup>1</sup>	Reserved										
0x53	XX <sup>1</sup>					served						
0x54	XX <sup>1</sup>			<del></del>	Re	served						

Value		Default					Bit #				
1.556	Addr	_	7	6	5	4	3	2	1	0	Description
1967   00	0x55	XX <sup>1</sup>				I. R	eserved	L			
1959   00	0x56	00		IP3[3:0]	_CFG4			RZ3[	3:0]_CFG4		PLL3 Loop Parameter
1955   00   19330   CFG2   19330   CFG3   19330	0x57	00		IP3[3:0]	_CFG5			RZ3[	3:0]_CFG5		
1955   1960   1973	0x58	00		IP3[3:0]	_CFG0			RZ3[	3:0]_CFG0		
1965   00	0x59	00									
Display   Dis											
0.55				IP3[3:0]	_CFG3						
March   Marc							٠ ,				
0.55E											
Digital   Digi											
Dagle   Dagl											
NST-PL-CFG4											
12-4096 (value of 1°0' through "1' are not allowed)   12-4096 (value of 1°0' through "1' are not of 1°			Reserved					i5			
No.60											
0x65   0C							<u>-</u>				
0x66         OC         N3[7:0]_CFG2           0x67         OC         N3[7:0]_CFG3           0x68         00         SSVCQ[7:0]_CFG1         Sepectrum Loop Feedback Counte Sep Addr (N860 x88 for Sep Addr (N860 x88) for Sep Add						-	-				
0x67						-	-				
DATE							<u>-</u>				_
Spectrum Loop Feedback Counte						\$\$VCO[7:0] PLL3 Sproad					
See Add 0.860.0x85 for   See See Add						Spectrum Loop Feedback Counter					
DATE											
Ox6C   Ox6D   Ox6C   Ox6D   Ox6C   Ox6D   Ox6C   Ox6D   Ox6C											SSVCO[15:8]
0x6D         00         SSVC0[7:0]_CFG5         SS_D3[7:0]_CFG4         SS_D[7:0]_FPL3 Spread Spectrum Meference Divide           0x6E         00         SS_D3[7:0]_CFG5         Reference Divide           0x70         00         SS_D3[7:0]_CFG1         Reference Divide           0x71         00         SS_D3[7:0]_CFG2         SS_D3[7:0]_CFG3           0x73         00         SS_D3[7:0]_CFG3         SS_D3[7:0]_CFG3           0x74         01         Reserved         S1         S3         Output Controls S1=1 - OUT1/OUT2 are from DIV1/DIV2 respectively S1=0 - Both from DIV2 S3=1 - OUT3/OUT2 are from DIV3/DIV6 S3=0 - Both from DIV6 OEM#—output enable mode x0 - instated o1 - park low 11 - pa											_
0x6E         00         SS_D3[7:0]_CFG4         SS_D[7:0] - PLL3 Spread Spectrum           0x6F         00         SS_D3[7:0]_CFG6         Reference Divide           0x70         00         SS_D3[7:0]_CFG0         Reference Divide           0x71         00         SS_D3[7:0]_CFG2         SS_D3[7:0]_CFG2           0x73         00         SS_D3[7:0]_CFG3         SS_D3[7:0]_CFG3           0x74         01         Reserved         S1         S3         Output Controls S1=1 - OUT1/OUT2 are from DIV1/DIV2 respectively S1=0 - Both from DIV2 S3=1 - OUT3/OUT6 are from DIV3/DIV6 S3=0 - Both from DIV2 S3=1 - OUT3/OUT6 are from DIV3/DIV6 S3=0 - Both from DIV6 OEM#-output enable mode x0 - tristated 01 - park low 11 - park low 111 - park low 02 DIV1/DIV2 DIV1/DI											_
Description											SS D[7:0] - PLL3 Spread Spectrum
0x70         00         SS_D3[7:0]_CFG0           0x71         00         SS_D3[7:0]_CFG1           0x72         00         SS_D3[7:0]_CFG2           0x73         00         SS_D3[7:0]_CFG3           0x74         01         Reserved           0x75         03         Reserved           0x76         03         Reserved           0x76         00         OEM1[1:0]         NV1[1:0]           0x76         00         OEM1[1:0]         NV1[1:0]           0x77         00         SLEW2[1:0]         Reserved           0x78         00         OEM3[1:0]         SLEW3[1:0]         Reserved           0x78         00         OEM3[1:0]         SLEW3[1:0]         INV3[1:0]         Reserved           0x78         00         OEM3[1:0]         SLEW5[1:0]         INV3[1:0]         Reserved           0x78         00         OEM5[1:0]         SLEW5[1:0]         INV5[1:0]         Reserved           0x7A         00         OEM5[1:0]         SLEW5[1:0]         Reserved											
0x71											_
0x73   00   SS_D3[7:0]_CFG3     0x74   01   Reserved     0x75   03   Reserved     0x76   00   SLEW3[1:0]   SLEW3[1:0]   SLEW5[1:0]   Reserved     0x77   00   SLEW3[1:0]   SLEW5[1:0]   Reserved     0x78   00   OEM5[1:0]   SLEW5[1:0]   Reserved     0x79   00   OEM5[1:0]   SLEW5[1:0]   Reserved     0x70   00   OEM5[1:0]   SLEW5[1:0]   Reserved     0x70   00   OEM5[1:0]   SLEW5[1:0]   Reserved     0x70   OEM5[1:0]   Reserved   OEM3 controls OUT3 and OUT6     0x70   00   OEM5[1:0]   SLEW5[1:0]   Reserved     0x70   OEM5[1:0]   Reserved   OEM5 controls OUT5 and OUT5 a	0x71	00									
DX74	0x72	00				SS_D3	3[7:0]_CFG2				
No.	0x73	00				SS_D3	3[7:0]_CFG3				
S1=1 - OUT1/OUT2 are from DIV1/DIV2 respectively S1=0 - Both from DIV2 S3 =1 - OUT3/OUT6 are from DIV3/DIV6 S3=0 - Both from DIV5 S3=0 - Both from DIV6 OEM#-output enable mode x0 - tristated 01 - park low 11 - park high    DX76	0x74	01				R	eserved				
INV1 [CLK1, CLK2]   [0] - normal   [1] - invert clock   OEM1 controls OUT1/OUT2	0x75	03								S1=1 - OUT1/OUT2 are from DIV1/DIV2 respectively S1=0 - Both from DIV2 S3 =1 - OUT3/OUT6 are from DIV3/DIV6 S3=0 - Both from DIV6 OEM#—output enable mode x0 - tristated 01 - park low	
0x78         00         OEM3[1:0]         SLEW3[1:0]         INV3[1:0]         Reserved         OEM3 controls OUT3 and OUT6           0x79         00         Reserved           0x7A         00         OEM5[1:0]         SLEW5[1:0]         INV5[1:0]         Reserved         OEM5 controls OUT5 and OUT5b           0x7B         00         SLEW6[1:0]         Reserved	0x76	00	OEM1	[1:0]			INV	[[1:0]	Re	served	INV1 [CLK1, CLK2] [0] - normal [1] - invert clock
0x79         00         Reserved           0x7A         00         OEM5[1:0]         SLEW5[1:0]         INV5[1:0]         Reserved         OEM5 controls OUT5 and OUT5b           0x7B         00         SLEW6[1:0]         Reserved         Reserved											
0x7A         00         OEM5[1:0]         SLEW5[1:0]         INV5[1:0]         Reserved         OEM5 controls OUT5 and OUT5b           0x7B         00         SLEW6[1:0]         Reserved										OEM3 controls OUT3 and OUT6	
0x7B 00 SLEW6[1:0] Reserved											
			OEM5	[1:0]			INV	o[1:0]			OEMS controls OUT5 and OUT5b
	0x7B 0x7C	00 XX <sup>1</sup>			SLEW		ocorrod	1	He	servea	

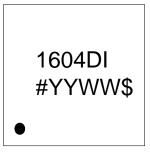
	Default				E	Bit #				
Addr	Register Hex Value	7	6	5	4	3	2	1	0	Description
0x7D	XX <sup>1</sup>			I	Re	eserved	1		L	
0x7E	XX <sup>1</sup>				Re	eserved				
0x7F	XX <sup>1</sup>				Re	eserved				
0x80	00				PLL3 Spread Spectrum Feedback					
0x81	00				SSVCO	[15:8]_CFG1				Counter
0x82	00				SSVCO	[15:8]_CFG2				
0x83	00				SSVCO	[15:8]_CFG3				
0x84	00				SSVCO	[15:8]_CFG4				
0x85	00				SSVCO	[15:8]_CFG5				
0x86	00				Re	eserved				
0x87	00				Re	eserved				
0x88	FF	PM1_CFG0				Q1[6:0]_CF0	GO			Output Divides
0x89	FF	PM1_CFG1				Q1[6:0]_CF0	31			for Q<>111111, —PM=0 - Divide by 2
A8x0	FF	PM1_CFG2				Q1[6:0]_CF0	G2			PM=1, (Q+2)*2
0x8B	FF	PM1_CFG3				Q1[6:0]_CF0	33			for Q=1111111
0x8C	FF	PM1_CFG4				Q1[6:0]_CF0	<b>3</b> 4			PM=0, disable the output divider PM=1, bypass the output divide,
0x8D	FF	PM1_CFG5				Q1[6:0]_CF0	<b>3</b> 5			(divide by 1)
0x8E	7F	PM2_CFG4				Q2[6:0]_CF0	34			
0x8F	7F	PM2_CFG5				Q2[6:0]_CF0	<b>3</b> 5			
0x90	7F	PM2_CFG0				Q2[6:0]_CF0	GO			
0x91	7F	PM2_CFG1				Q2[6:0]_CF0	31			
0x92	7F	PM2_CFG2				Q2[6:0]_CF0	32			
0x93	7F	PM2_CFG3				Q2[6:0]_CF0	G3			
0x94	7F	PM3_CFG0				Q3[6:0]_CF0	GO			
0x95	7F	PM3_CFG1				Q3[6:0]_CF0	31			
0x96	7F	PM3_CFG2				Q3[6:0]_CF0				
0x97	7F	PM3_CFG3				Q3[6:0]_CF0	33			
0x98	7F	PM3_CFG4				Q3[6:0]_CF0				
0x99	7F	PM3_CFG5				Q3[6:0]_CF0				
0x9A	7F	PM4_CFG4				Q4[6:0]_CF0				
0x9B	7F	PM4_CFG5				Q4[6:0]_CF0				
0x9C	7F	PM4_CFG0				Q4[6:0]_CF0				
0x9D	7F	PM4_CFG1				Q4[6:0]_CF0				
0x9E	7F	PM4_CFG2				Q4[6:0]_CF0				
0x9F	7F	PM4_CFG3				Q4[6:0]_CF0				
0xA0	7F	PM5_CFG0				Q5[6:0]_CF0				
0xA1	7F	PM5_CFG1				Q5[6:0]_CF0				
0xA2	7F	PM5_CFG2				Q5[6:0]_CF0				
0xA3	7F	PM5_CFG3				Q5[6:0]_CF0				
0xA4	7F	PM5_CFG4								
0xA5	7F	PM5_CFG5	Q5[6:0]_CFG5 Q6[6:0]_CFG4							
0xA6	7F	PM6_CFG4								
0xA7	7F	PM6_CFG5				Q6[6:0]_CF0				
0xA8	7F	PM6_CFG0								
0xA9	7F	PM6_CFG1								
0xAA	7F	PM6_CFG2				Q6[6:0]_CF0				
0xAB	7F	PM6_CFG3				Q6[6:0]_CF0	à3			

	Default					Bit #				
Addr	Register Hex Value	7	6	5	4	3	2	1	0	Description
0xAC	00		TSSC[3:	0]_CFG0	1	I.	NSSC[	- I	PLL0 Spread Spectrum Control	
0xAD	00		TSSC[3:	0]_CFG1			NSSC[			
0xAE	00		TSSC[3:0]_CFG2				NSSC[			
0xAF	00	TSSC[3:0]_CFG3					NSSC[	3:0]_CFG3		
0xB0	00	TSSC[3:0]_CFG4					NSSC[	3:0]_CFG4		
0xB1	00		TSSC[3:	0]_CFG5			NSSC[	3:0]_CFG5		
0xB2	00	DITH_CFG4	X2_CFG4			SSOFFSE	T[5:0]_CFG4			
0xB3	00	DITH_CFG5	X2_CFG5			SSOFFSE	T[5:0]_CFG5			
0xB4	00	DITH_CFG0	X2_CFG0			SSOFFSE	T[5:0]_CFG0			
0xB5	00	DITH_CFG1	X2_CFG1			SSOFFSE	T[5:0]_CFG1			
0xB6	00	DITH_CFG2	X2_CFG2			SSOFFSE	T[5:0]_CFG2			
0xB7	00	DITH_CFG3	X2_CFG3			SSOFFSE	T[5:0]_CFG3			
0xB8	11		SD1[3:0	)]_CFG0			SD0[3	:0]_CFG0		
0xB9	11		SD1[3:0	)]_CFG1			SD0[3	:0]_CFG1		
0xBA	11		SD1[3:0	)]_CFG2			SD0[3	:0]_CFG2		
0xBB	11		SD1[3:0	)]_CFG3			SD0[3	:0]_CFG3		
0xBC	11		SD1[3:0	)]_CFG4			SD0[3	:0]_CFG4		
0xBD	11	SD1[3:0]_CFG5					SD0[3	:0]_CFG5		
0xBE	AE	SRC1[1:0	D]_CFG4			PDPL3_CFG4	SM[1:0	]_CFG4	PRIMSRC_CFG4	Output Divide Source Selection
0xBF	AE	SRC1[1:0	)]_CFG5			PDPL3_CFG5	SM[1:0	]_CFG5	PRIMSRC_CFG5	PRIMSRC - primary source - crystal or ICLOCK 0 = crystal/REFIN 1 = CLKIN
0xC0	AE	SRC1[1:0	)]_CFG0			PDPL3_CFG0	PDPL3_CFG0 SM[1:0]_CFG0 PRIMSRC_CFG0		PRIMSRC_CFG0	SM = switch mode 0x = manual 10 = reserved 11 = auto-revertive
0xC1	AE	SRC1[1:0	)]_CFG1			PDPL3_CFG1	3_CFG1 SM[1:0]_CFG1 PRIMSRC_C		PRIMSRC_CFG1	PDPL3 - PLL3 shutdown 0 = normal 1 = shut down
0xC2	AE	SRC1[1:0	)]_CFG2			PDPL3_CFG2	SM[1:0	]_CFG2	PRIMSRC_CFG2	SRC = MUX control bit prior to DIV# SRC0[1:0] 00 - DIV1 01 - DIV3 10 - Reference input
0xC3	AE	SRC1[1:0	D]_CFG3			PDPL3_CFG3	SM[1:0	]_CFG3	PRIMSRC_CFG3	
0xC4	24		:	SRC3[2:0]_CFG	0	S	RC2[2:0]_CFG	0	SRC1[2]_CFG0	SRC1/SRC2/SRC3SRC5
0xC5	24		;	SRC3[2:0]_CFG	1	S	RC2[2:0]_CFG	1	SRC1[2]_CFG1	000 - DIV1 001 - DIV3
0xC6	24			SRC3[2:0]_CFG	2	S	RC2[2:0]_CFG	2	SRC1[2]_CFG2	010 - DIV3
0xC7	24		;	SRC3[2:0]_CFG	3	S	RC2[2:0]_CFG	3		011 - Reserved
0xC8	24		;	SRC3[2:0]_CFG	4	S	RC2[2:0]_CFG	4	SRC1[2]_CFG4	100 - PLL0 101 - PLL1
0xC9	24		,	SRC3[2:0]_CFG	5	S	RC2[2:0]_CFG	5	SRC1[2]_CFG5	110 - PLL2 111 - PLL3
0xCA	49	S	RC6[2:0]_CFG	i4		SRC5[2:0]_CFG4				SRC6
0xCB	49	S	RC6[2:0]_CFG	ì5		SRC5[2:0]_CFG5		Ì		000 - Reserved
0xCC	49	S	RC6[2:0]_CFG	i0						001 - Reserved 010 - Reference input
0xCD	49	S				1 SRC5[2:0]_CFG1				011 - Reserved
0xCE	49		RC6[2:0]_CFG			SRC5[2:0]_CFG2				100 - Reserved 101 - PLL1
0xCF	49	S	RC6[2:0]_CFG	<del>3</del> 3		SRC5[2:0]_CFG3				110 - Reserved 111 - Reserved Quiet MUX

Default Configuration: OUT1 = Reference Clock output, all other outputs turned off.

<sup>&</sup>lt;sup>1</sup>. Memory bytes do not exist. Readback will be last value in shift register. If reading sequentially, value in 0x51 will be returned.

# **Marking Diagram**



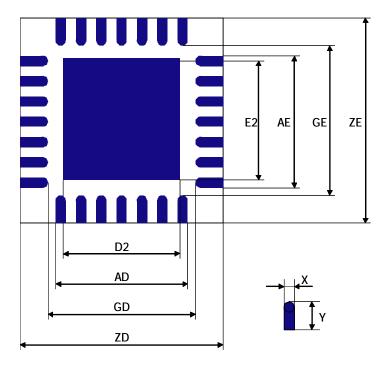
#### Notes:

- 1. "#" is the lot number.
- 2. YYWW is the last two digits of the year and week that the part was assembled.
- 3. "\$" is the assembly mark code.
- 4. "I" indicates industrial temperature range.

# **Thermal Characteristics 28-pin QFN**

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Units
Thermal Resistance Junction to	$\theta_{JA}$	Still air		47.0		° C/W
Ambient	$\theta_{JA}$	1 m/s air flow		41.8		° C/W
	$\theta_{JA}$	2.5 m/s air flow		39.2		° C/W
Thermal Resistance Junction to Case	$\theta_{JC}$			52.9		° C/W

# **Landing Pattern**

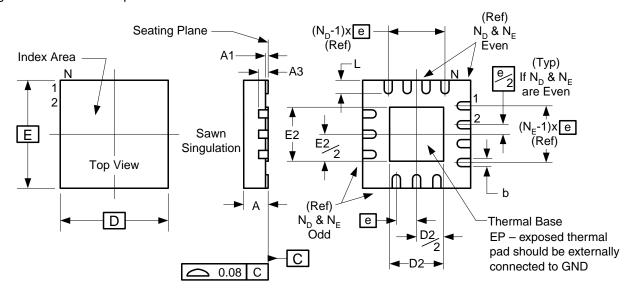


Dimensions				
X(max)	0.25			
Yref	0.76			
A(max)	2.65			
G(min)	2.9			
Z(max)	4.41			
E2/D2(max)	2.7			

Unit: mm

### Package Outline and Package Dimensions (28-pin 4mm x 4mm QFN)

Package dimensions are kept current with JEDEC Publication No. 95



	Willin	neters		
Symbol	Min	Max		
Α	0.80	1.00		
A1	0	0.05		
A3	0.20 Reference			
b	0.15	0.25		
е	0.40 E	BASIC		
N	2	28		
N <sub>D</sub>	•	7		
N <sub>E</sub>		7		
D x E BASIC	4.00	x 4.00		
D2	2.50	2.70		
E2	2.50	2.70		
L	0.30	0.50		

# **Ordering Information**

Part / Order Number	Marking	<b>Shipping Packaging</b>	Package	Temperature
5V19EE604NDGI	See Page 26	Trays	28-pin QFN	-40 to +85° C
5V19EE604NDGI8	See Page 26	Tape and Reel	28-pin QFN	-40 to +85° C

#### "G" after the two-letter package code are the Pb-Free configuration and are RoHS compliant.

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# **Revision History**

Rev.	Date	Originator	Description of Change
Α	4/22/09	R.Willner	Advance Information.
В	5/04/09	R.Willner	Identified VDDX (crystal oscillator power) and AVDD (analog power) on device.
С	6/04/09	R.Willner	Add default configurations, pull-down resistor values on input pins. Released Datasheet from Advanced Information.
D	06/10/09	R.Willner	Updates: crystal load specs; "Output Duty Cycle" specs; addresses 0x07, 0x02 and 0xBF in "Programming Registers" table.
Е	10/05/09	R.Willner	Change IP3[3:0] to IP3[4:0]; updated "Programming Registers Table".
F	02/23/10	R.Willner	Updated Recommended Operation Conditions to include Vddx and AVdd parameters.
G	01/19/11	R.Willner	Corrected notes for top-side marking.
Н	05/04/11	R.Willner	Added Landing Pattern diagram.
J	04/18/12	R. Willner	<ol> <li>Change description for SDAT and SCLK pins.</li> <li>Add new footnotes to pin descriptions table</li> <li>Added section "Crystal Clock Selection"</li> <li>Added logic diagram and Truth table for "SD/OE Pin Function" section.</li> <li>Corrected register readback values for 0x52~0x54 and 0x7C~0x7F.</li> <li>Update to QFN package drawing - exposed thermal pad callout.</li> </ol>
К	06/04/12	A. Tsui	Updated SD-OE pin description; from (Default is active HIGH) to (Default is active LOW)     Updated "OUTn" column in Truth Table with "High-Z" specs and added footnote 2, "High-Z regardless of OEM bits".     Updated "SD-OE Pin Function" section to reflect that SP is "0"changed from active HIGH to active LOW, and SP is "1" changed from active LOW to active HIGH.
L	06/18/12	R.Willner	Added Min/Max spread values to "Spread Spectrum Generation Specifications" table; fMOD - Max. 120kHz; Down Spread - Min0.5%, Max4.0%; Center Spread - Min. ±0.25%, Max. ±2.0%
М	09/24/12	R.Willner	Slew Rate (t4) Output Load test conditions were changed from 15pF to 5pF.

IDT5V19EE604 EEPROM PROGRAMMABLE VCXO CLOCK GENERATOR

**CLOCK SYNTHESIZER** 

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